

SG6859D

Low-Cost, Green-Mode PWM Controller for Flyback Converters

Features


- Green-Mode PWM
- Supports the “Blue Angel” Standard
- Low Startup Current: 9µA
- Low Operating Current: 3mA
- 300mA Driving Capability
- Leading-Edge Blanking
- Constant Output Power Limit
- Universal Input
- Built-in Synchronized Slope Compensation
- Current-Mode Operation
- Cycle-by-Cycle Current Limiting
- Under-Voltage Lockout (UVLO)
- Programmable PWM Frequency with Frequency Hopping
- V_{DD} Over-Voltage Protection (Auto Restart)
- Gate Output Voltage Clamped at 17V
- Low Cost
- Few External Components Required
- Small SSOT-6 Package

Applications

General-purpose switching mode power supplies and flyback power converters, such as:

- Battery chargers for cellular phones, cordless phones, PDAs, digital cameras, and power tools
- Power adapters for ink jet printers, video game consoles, and portable audio players
- Open-frame SMPS for TV/DVD standby and auxiliary supplies, home appliances, and consumer electronics
- Replacements for linear transformers and RCC SMPS
- PC 5V standby power

Ordering Information

Part Number	Operating Temperature Range	Package	 Eco Status	Packing Method
SG6859DTY	-40 to +105°C	SSOT-6	Green	Tape & reel
SG6859DDY		DIP-8		Tube

 For Fairchild’s definition of “green” Eco Status, please visit: http://www.fairchildsemi.com/company/green/rohs_green.html.

Description

This highly integrated PWM controller provides several enhancements designed to meet the low standby-power needs of low-power SMPS. To minimize standby power consumption, the proprietary green-mode function provides off-time modulation to linearly decrease the switching frequency under light-load conditions. This green-mode function enables the power supply to meet even the strictest power conservation requirements.

The BiCMOS fabrication process enables reducing the startup current to 9µA and the operating current to 3mA. To further improve power conservation, a large startup resistance can be used. Built-in synchronized slope compensation ensures the stability of peak current mode control. Proprietary internal compensation provides a constant output power limit over a universal AC input range (90V_{AC} to 264V_{AC}). Pulse-by-pulse current limiting ensures safe operation during short-circuits.

To protect the external power MOSFET from damage by supply over voltage, the SG6859D’s output driver is clamped at 17V. SG6859D controllers can be used to improve the performance and reduce the production cost of power supplies. The best choice for replacing linear and RCC-mode power adapters. Available in 8-pin DIP and 6-pin SSOT-6 packages.

Application Diagram

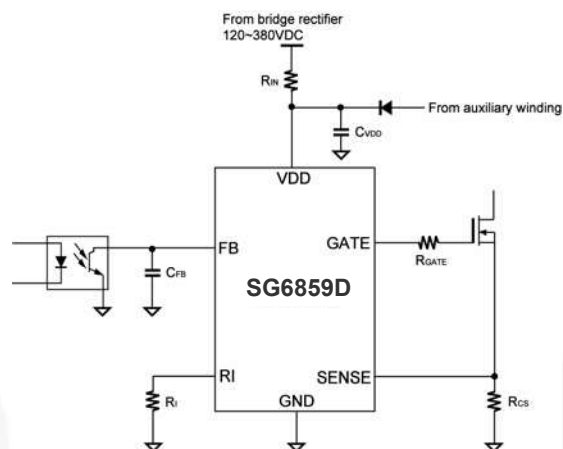


Figure 1. Typical Application

Internal Block Diagram

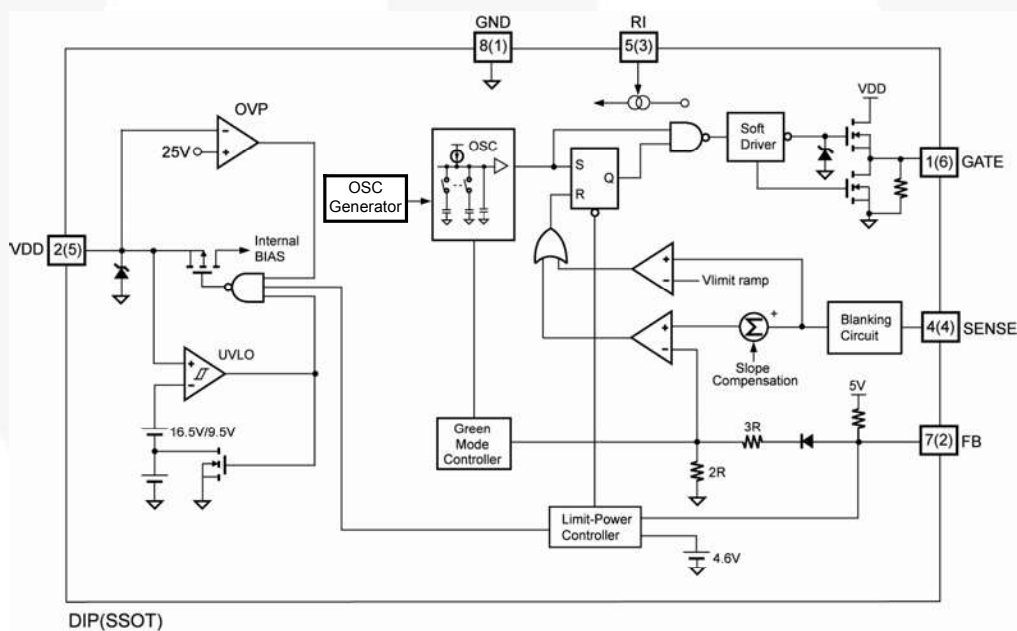
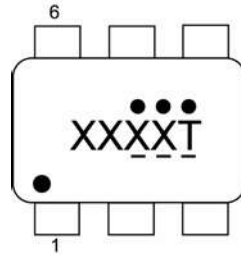


Figure 2. Functional Block Diagram

Marking Information



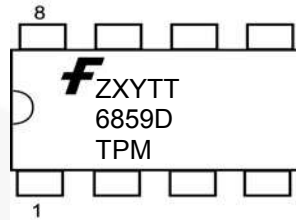
XXXX: ABBK=SG6859D

T : Die Run Code

. . . : Year Code

- - - : Week Code

*Marking for SG6859DTY (Green compound)



F- Fairchild Logo

Z- Plant Code

X- 1 Digit Year Code

Y- 1 Digit Week Code

TT: 2 Digits die Run Code

T: Package Type (D=DIP)

P: Y: Green Package

M: Manufacture Flow Code

*Marking for SG6859DDY (Green compound)

Figure 3. Top Mark

Pin Configurations

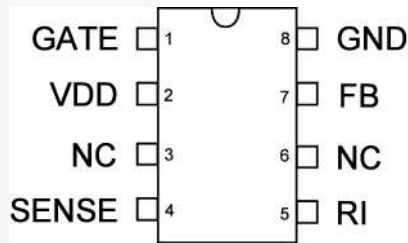


Figure 4. 8-Pin DIP

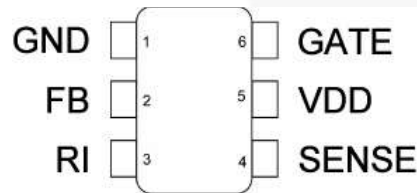


Figure 5. 6-Pin SSOT

Pin Definitions

DIP Pin #	SSOT Pin #	Name	Description
1	6	GATE	The totem-pole output driver for driving the power MOSFET.
2	5	VDD	Power Supply.
3		NC	No Connection.
4	4	SENSE	Current Sense. This pin senses the voltage across a resistor. When the voltage reaches the internal threshold, PWM output is disabled. This activates over-current protection. This pin provides current amplitude information for current-mode control.
5	3	RI	A resistor connected from the RI pin to ground generates a constant current source used to charge an internal capacitor and determine the switching frequency. Increasing the resistance reduces the amplitude of the current source and the switching frequency. A 95kΩ resistor R _i results in a 50μA constant current I _i and a 65kHz switching frequency.
6		NC	No Connection.
7	2	FB	Feedback. The FB pin provides the output voltage regulation signal. It provides feedback to the internal PWM comparator, so that the PWM comparator can control the duty cycle.
8	1	GND	Ground.

Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

Symbol	Parameter		Min.	Max.	Unit
V _{VDD}	DC Supply Voltage ^(1, 2)			30	V
V _{FB}	Input Voltage to FB Pin		-0.3	7.0	V
V _{SENSE}	Input Voltage to Sense Pin		-0.3	7.0	V
T _J	Operating Junction Temperature			150	°C
Θ _{JA}	Thermal Resistance (Junction-to-Air)	SSOT		208.4	°C/W
		DIP		82.5	°C/W
T _{STG}	Storage Temperature Range		-55	+150	°C
T _L	Lead Temperature (Wave Soldering or IR, 10 Seconds)			+260	°C
ESD	Electrostatic Discharge	Human Body Model, JEDEC:JESD22-A114		3.5	kV
		Charged Device Model, JEDEC:JESD22-C101		1.25	kV

Notes:

1. All voltage values, except differential voltages, are given with respect to GND pin.
2. Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device.

Electrical Characteristics

Unless otherwise noted, $V_{DD}=15V$ and $T_A=25^{\circ}C$.

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units	
V _{DD} Section							
V _{DD-OP}	Continuously Operation Voltage				22	V	
V _{DD-ON}	Turn-On Threshold Voltage		15.5	16.5	17.5	V	
V _{DD-OFF}	Turn-Off Threshold Voltage		8.5	9.5	10.5	V	
I _{DD-ST}	Startup Current	V _{DD} =V _{DD-ON} – 0.1V		9	15	μA	
I _{DD-OP}	Operating Supply Current	V _{DD} =15V, C _L =1nF		3.0	3.5	mA	
V _{DD-OVP}	V _{DD} Over-Voltage Protection Level	Auto Restart	24	25	26	V	
t _{D-VDDOVP}	V _{DD} Over-Voltage Protection Debounce	Auto Restart		125		μs	
Feedback Input Section							
Z _{FB}	Input Impedance			5		kΩ	
V _{FB-OPEN}	FB Output High Voltage	FB Pin Open	5			V	
V _{FB-OL}	FB Open-loop Trigger Level		4.3	4.6	4.9	V	
t _{D-OLP}	Delay of FB Pin Open-loop Protection	R _I =95kΩ	53	58	63	ms	
V _{FB-N}	Green-Mode Entry FB Voltage		2.60	2.85	3.10	V	
V _{FB-G}	Green-Mode Ending FB Voltage			2.2		V	
V _{FB-ZDC}	Zero Duty-Cycle Input Voltage			1.8		V	
S _G	Green-Mode Modulation Slope	R _I =95kΩ	40	60	80	Hz/mV	
Current-Sense Section							
Z _{SENSE}	Input Impedance		10			kΩ	
t _{PD}	Delay to Output		45	65	85	ns	
V _{STHFL}	Flat Threshold Voltage for Current Limit		0.95	1.01	1.06	V	
V _{STHVA}	Valley Threshold Voltage for Current Limit		0.75	0.80	0.85	V	
t _{LEB}	Leading-Edge Blanking Time		250	300	350	ns	
DCY _{SAW}	Duty Cycle of SAW Limit			30		%	
Oscillator Section							
f _{OSC}	Frequency	Center Frequency	R _I =95kΩ	60	65	70	kHz
		Hopping Range			±4		
t _{HOP}	Hopping Period	R _I =95kΩ		4		ms	
f _{OSC-G}	Green-Mode Frequency	R _I =95kΩ		22		kHz	
f _{DV}	Frequency Variation vs. V _{DD} Deviation	V _{DD} =13.5 to 22V			5	%	
f _{DT}	Frequency Variation vs. Temperature Deviation	T _A =-20 to 85°C			5	%	
Output Section							
DCY _{MAX}	Maximum Duty Cycle		65	70	75	%	
V _{GATE-L}	Output Voltage Low	V _{DD} =15V, I _O =20mA			1.4	V	
V _{GATE-H}	Output Voltage High	V _{DD} =13.5V, I _O =20mA	8			V	
t _r	Rising Time	V _{DD} =15V, C _L =1nF	80	150	180	ns	
t _f	Falling Time	V _{DD} =15V, C _L =1nF	35	55	75	ns	
V _{GATE-CLAMP}	Output Clamp Voltage	V _{DD} =22V	16	17	18	V	

Typical Performance Characteristics

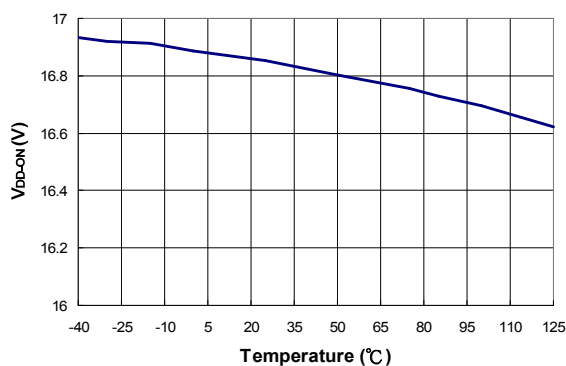


Figure 6. Turn-On Threshold Voltage (V_{DD-ON}) vs. Temperature

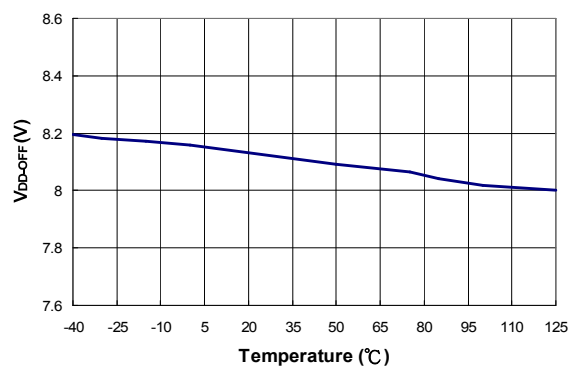


Figure 7. Turn-Off Threshold Voltage (V_{DD-OFF}) vs. Temperature

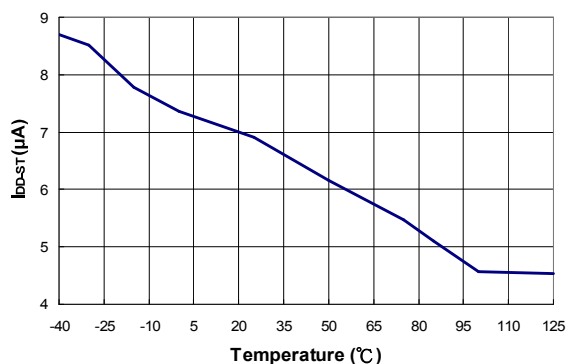


Figure 8. Startup Current (I_{DD-ST}) vs. Temperature

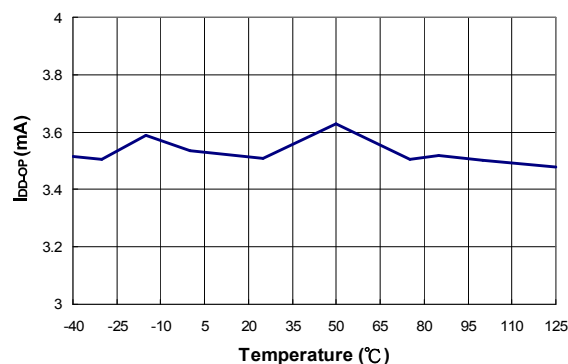


Figure 9. Operating Supply Current (I_{DD-OP}) vs. Temperature

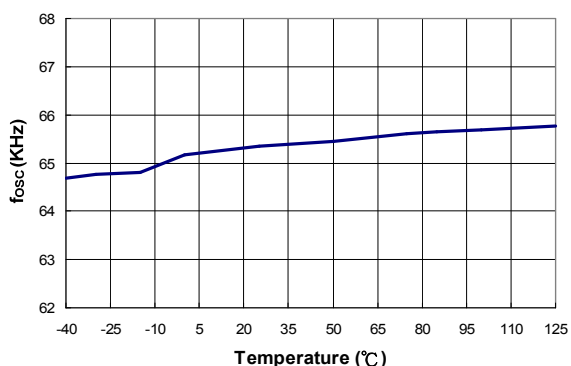


Figure 10. Center Frequency (f_{OSC}) vs. Temperature

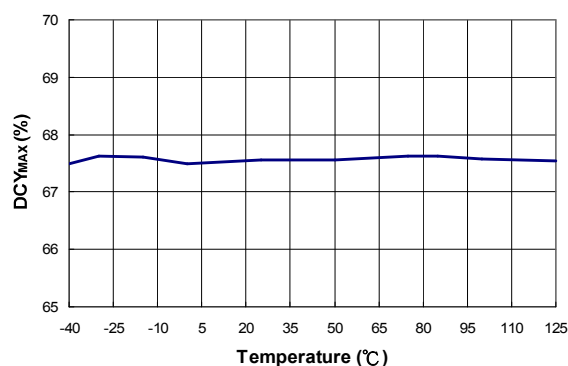


Figure 11. Maximum Duty Cycle (DCY_{MAX}) vs. Temperature

Typical Performance Characteristics

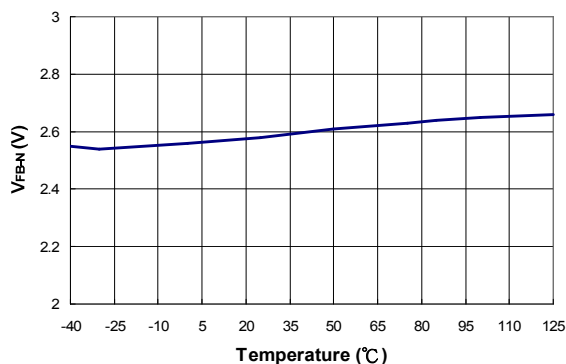


Figure 12. Green-Mode Entry FB Voltage (V_{FB-N}) vs. Temperature

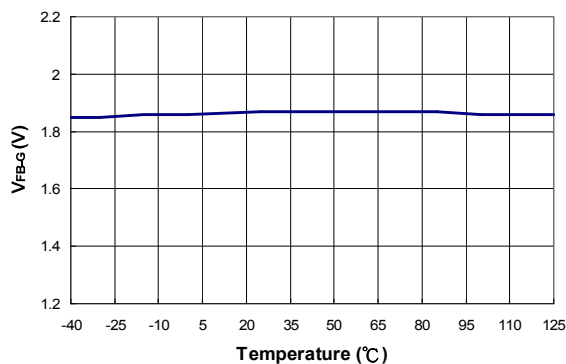


Figure 13. Green-Mode Ending FB Voltage (V_{FB-G}) vs. Temperature

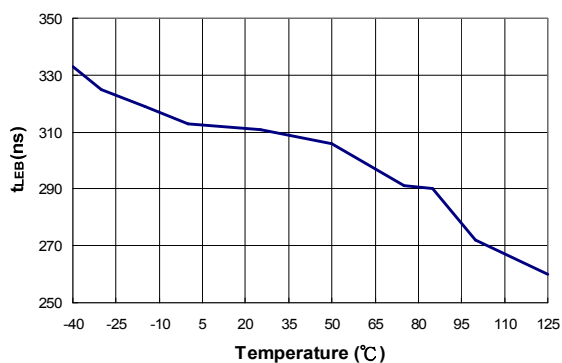


Figure 14. Leading-Edge Blanking Time (t_{LEB}) vs. Temperature

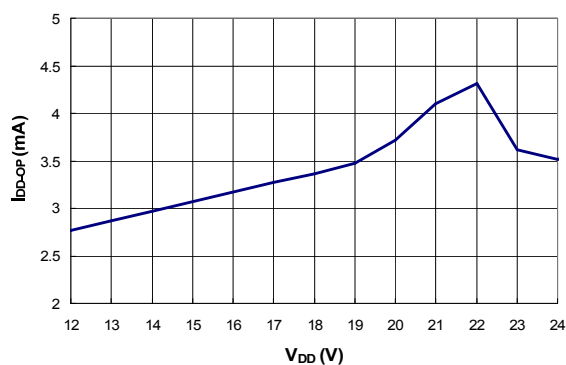


Figure 15. Operating Supply Current vs. V_{DD} Voltage

Operation Description

SG6859D devices integrate many useful designs into one controller for low-power, switch-mode power supplies. The following descriptions highlight some of the features of the SG6859D series.

Startup Current

The startup current is only 9μA. Low startup current allows a startup resistor with high resistance and low wattage to supply the startup power for the controller. A 1.5MΩ, 0.25W, startup resistor and a 10μF/25V V_{DD} hold-up capacitor are sufficient for an AC-to-DC power adapter with a wide input range (100V_{AC} to 240V_{AC}).

Operating Current

The operating current has been reduced to 3mA. The low operating current results in higher efficiency and reduces the V_{DD} hold-up capacitance requirement.

Green-Mode Operation

The proprietary green-mode function provides off-time modulation to linearly decrease the switching frequency under light-load conditions. On-time is limited to provide stronger protection against brownouts and other abnormal conditions. The feedback current, which is sampled from the voltage feedback loop, is taken as the reference. Once the feedback current exceeds the threshold current, the switching frequency starts to decrease. This green-mode function dramatically reduces power consumption under light-load and zero-load conditions. Power supplies using the SG6859D can meet even the strictest regulations regarding standby power consumption.

Oscillator Operation

A resistor connected from the RI pin to ground generates a constant current source used to charge an internal capacitor. The charge time determines the internal clock speed and the switching frequency. Increasing the resistance reduces the amplitude of the input current and the switching frequency. A 95kΩ R_i resistor results in a 50μA constant current I_i and a 65kHz switching frequency. The relationship between R_i and the switching frequency is:

$$f_{\text{PWM}} = \frac{6175}{R_i(\text{k}\Omega)} (\text{kHz}) \quad (1)$$

Leading-Edge Blanking (LEB)

Each time the power MOSFET is switched on, a turn-on spike occurs at the sense-resistor. To avoid premature termination of the switching pulse, a 300ns leading-edge blanking time is built in. Conventional RC filtering can be omitted. During this blanking period, the current-limit comparator is disabled and cannot switch off the gate driver.

Constant Output Power Limit

When the SENSE voltage across the sense resistor, R_S, reaches the threshold voltage (around 1.01V), the output GATE drive is turned off following a short propagation delay, t_{PD}. This propagation delay introduces an additional current proportional to t_{PD}·V_{IN}/L_P. The propagation delay is nearly constant, regardless of the input line voltage V_{IN}. Higher input line voltages result in larger additional currents. At high input line voltages, the output power limit is higher than at low input line voltages. To compensate for this output power limit variation across a wide AC input range, the threshold voltage is adjusted by adding a positive ramp. This ramp signal rises from 0.8V to 1.01V, then flattens out at 1.01V. A smaller threshold voltage forces the output GATE drive to terminate earlier, which reduces the total PWM turn-on time and makes the output power equal to that of low line input. This proprietary internal compensation ensures a constant output power limit for a wide AC input voltage range (90V_{AC} to 264V_{AC}).

Under-Voltage Lockout (UVLO)

The turn-on and -off thresholds are fixed internally at 16.5V and 9.5V. During startup, the hold-up capacitor must be charged to 16.5V through the startup resistor to enable the SG6859D. The hold-up capacitor continues to supply V_{DD} until power is delivered from the auxiliary winding of the main transformer. V_{DD} must not drop below 9.5V during this startup process. This UVLO hysteresis window ensures that hold-up capacitor is adequate to supply V_{DD} during startup.

Gate Output

The BiCMOS output stage is a fast totem pole gate driver. Cross conduction has been avoided to minimize heat dissipation, increase efficiency, and enhance reliability. The output driver is clamped by an internal 17V Zener diode to protect power MOSFET transistors against undesired over-voltage gate signals.

Built-In Slope Compensation

The sensed voltage across the current-sense resistor is used for current-mode control and pulse-by-pulse current limiting. Built-in slope compensation improves stability and prevents sub-harmonic oscillations due to peak-current mode control. The SG6859D has a synchronized, positively-sloped ramp built-in at each switching cycle.






Noise Immunity

Noise from the current sense or the control signal can cause significant pulse-width jitter, particularly in continuous-conduction mode (CCM). While slope compensation helps alleviate these problems, further precautions should still be taken. Good placement and layout practices should be followed. Avoiding long PCB traces and component leads, locating compensation and filter components near the SG6859D, and increasing power MOS gate resistance improve performance.



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Definition of Terms

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