## FMS3110／FMS3115 <br> Triple Video D／A Converters，3x10－Bit，150Ms／s

## Features

－10－bit Resolution
－ 150 Megapixels per Second
－$\pm 0.1 \%$ Linearity Error
－ISYNC and／BLANK Controls
－ $1.0 \mathrm{~V}_{\mathrm{PP}}$ Video into $37.5 \Omega$ or $75 \Omega$ Load
－Internal Bandgap Voltage Reference
－Double－buffered Data for Low Distortion
－TTL－compatible Inputs
－Low Glitch Energy
－Single＋5V Power Supply

## Applications

－Video Signal Conversion
－RGB
$-\mathrm{YC}_{B} \mathrm{C}_{\mathrm{R}}$
－Composite，Y，C
－Multimedia Systems
－Image Processing
－True－color Graphics Systems： 1 Billion Colors
－Broadcast Television Equipment
－High－Definition Television（HDTV）Equipment
－Direct Digital Synthesis

## Description

FMS3110／FMS3115 products are low－cost，triple D／A converters tailored for graphics and video applications where speed is critical．Two speed grades are available：
－FMS3110－100Ms／s
－FMS3115－150Ms／s
TTL－level inputs are converted to analog current outputs that can drive 25－37．5 $\Omega$ loads corresponding to doubly terminated $50-75 \Omega$ loads．A sync current following ISYNC input timing is added to the log output．／BLANK overrides RGB inputs，setting log，$_{\text {Ob，}}$ and Ior currents to zero when／BLANK $=$ L．Although appropriate for many applications，the internal 1.235 V reference voltage can be overridden by the $\mathrm{V}_{\text {REF }}$ input．
Few external components are required：a current reference resistor，current output load resistors，and decoupling capacitors．
Package is a 48－lead LQFP．Fabrication technology is CMOS．Performance is guaranteed from 0 to $70^{\circ} \mathrm{C}$ ．


Figure 1．Block Diagram

## Ordering Information

| Part <br> Numbers | Conversion <br> Rate | Operating <br> Temperature Range | Screening | Package |
| :---: | :---: | :---: | :---: | :---: |
| FMS3110KRC | $100 \mathrm{Ms} / \mathrm{s}$ | 0 to $70^{\circ} \mathrm{C}$ | Commercial | 48－Contact Leadless Quad Flat Package |
| FMS3115KRC | $150 \mathrm{Ms} / \mathrm{s}$ | 0 to $70^{\circ} \mathrm{C}$ | Commercial | 48－Contact Leadless Quad Flat Package |

All packages are lead free per JEDEC：J－STD－020B standard．

Pin Definitions

| Name | Pin \# | Value | Description |
| :---: | :---: | :---: | :---: |
| Clock / Pixel I/O |  |  |  |
| CLK | 26 | TTL | Clock Input. The clock input is TTL-compatible and all pixel data is registered on the rising edge of CLK. It is recommended that CLK be driven by a dedicated TTL buffer to avoid reflection-induced jitter, overshoot, and undershoot. |
| R9-0 | 47-37 | TTL | Red Pixel Data Inputs. TTL-compatible red data inputs are registered on the rising edge of CLK. |
| G9-0 | 48, 9-1 | TTL | Green Pixel Data Inputs. TTL-compatible green data inputs are registered on the rising edge of CLK. |
| B9-0 | 23-14 | TTL | Blue Pixel Data Inputs. TTL-compatible blue data inputs are registered on the rising edge of CLK. |
| Controls |  |  |  |
| ISYNC | 11 | TTL | Sync Pulse Input. Bringing SYNC LOW turns off a 40 IRE ( 7.62 mA ) current source which forms a sync pulse on the green D/A converter output. /SYNC is registered on the rising edge of CLK with the same pipeline latency as /BLANK and pixel data. ISYNC does not override any other data and should be used only during the blanking interval. <br> Since this is a single-supply D/A and all signals are positive-going, /SYNC is added to the bottom of the green D/A range. Turning /SYNC OFF means turning the current source ON. When a sync pulse is desired, the current source is turned OFF. If the system does not require sync pulses from the green D/A converter, /SYNC should be connected to GND. |
| /BLANK | 10 | TTL | Blanking Input. When /BLANK is LOW, pixel inputs are ignored and the D/A converter outputs fall to the blanking level. /BLANK is registered on the rising edge of CLK and has the same pipeline latency as /SYNC. |
| Video Outputs |  |  |  |
| $1 \mathrm{O}_{\mathrm{R}}$ | 33 | $0.714 \mathrm{~V}_{\mathrm{pp}}$ | Red Current Output. The current source outputs of the D/A converters are capable of driving RS-343A/SMPTE-170M-compatible levels into doubly terminated $75 \Omega$ lines. |
| $1 \mathrm{O}_{\mathrm{G}}$ | 32 | $1 \mathrm{~V}_{\text {PP }}$ | Green Current Output. The current source outputs of the D/A converters are capable of driving RS-343A/SMPTE-170M-compatible levels into doubly terminated $75 \Omega$ lines. Sync pulses may be added to the green D/A output. |
| IOB | 29 | $0.714 \mathrm{~V}_{\mathrm{PP}}$ | Blue Current Output. The current source outputs of the D/A converters are capable of driving RS-343A/SMPTE-170M-compatible levels into doubly terminated $75 \Omega$ lines. |
| Voltage Reference |  |  |  |
| $V_{\text {REF }}$ | 35 | +1.235V | Voltage Reference Output/Input. An internal voltage source of +1.235 V is output on this pin. An external +1.235 V reference may be applied here to override the internal reference. Decoupling VREF to GND with a $0.1 \mu \mathrm{~F}$ ceramic capacitor is required. |
| $V_{\text {ReF }}$ | 36 | $560 \Omega$ | Current-Setting Resistor. Full-scale output current of each D/A converter is determined by the value of the resistor connected between R $_{\text {REF }}$ and GND. Nominal value of $R_{\text {REF }}$ is found from: $R_{R E F}=9.1\left(V_{R E F} / I_{F S}\right)$ <br> where $I_{\text {Fs }}$ is the full-scale (white) output current (in amps) from the D/A converter (without sync). Sync is $0.4 \cdot I_{\text {Fs }}$. <br> D/A full-scale (white) current may also be calculated from: $\mathrm{I}_{\mathrm{FS}}=\mathrm{V}_{\mathrm{FS}} / \mathrm{R}_{\mathrm{L}}$ <br> where $V_{F S}$ is the white voltage level and $R_{L}$ is the total resistive load (in $\Omega$ ) on each D/A converter. $V_{F S}$ is the blank to full-scale voltage. |
| COMP | 34 | $0.1 \mu \mathrm{~F}$ | Compensation Capacitor. A $0.1 \mu \mathrm{~F}$ ceramic capacitor must be connected between COMP and $V_{D D}$ to stabilize internal bias circuitry. |
| Power and Ground |  |  |  |
| $V_{D D}$ | 12, 30, 31 | +5V | Power Supply. |
| GND | 27, 28 | 0.0 V | Ground. |

## Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only. See Figure 4.

| Symbol | Parameter | Min. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: |
| Power Supply Voltage |  |  |  |  |
| $V_{\text {DD }}$ | Measured to Ground | -0.5 | 7.0 | V |
| Inputs |  |  |  |  |
| VIN_A | Applied Voltage, Measured to Ground ${ }^{(1)}$ | -0.5 | $\mathrm{V}_{\mathrm{DD}}+0.5$ | V |
| lin_F | Forced Current ${ }^{(2,3)}$ | -10 | +10 | mA |
| Outputs |  |  |  |  |
| Vout_A | Applied Voltage, Measured to Ground ${ }^{(1)}$ | -0.5 | $\mathrm{V}_{\mathrm{DD}}+0.5$ | V |
| lout_f | Forced Current ${ }^{(2,3)}$ | -60 | +60 | mA |
| $\mathrm{t}_{\mathrm{sc}}$ | Short-Circuit Duration, Single Output in HIGH State to Ground |  | Infinite | S |
| Temperature |  |  |  |  |
| $\mathrm{T}_{\text {A }}$ | Operating Ambient Temperature | -20 | +110 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\text {J }}$ | Junction Temperature |  | +150 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{L}}$ | Lead Soldering, 10 Seconds |  | +300 | ${ }^{\circ} \mathrm{C}$ |
| TVP | Vapor Phase Soldering, 1 Minute |  | +220 | ${ }^{\circ} \mathrm{C}$ |
| TSTG | Storage Temperature | -65 | +150 | ${ }^{\circ} \mathrm{C}$ |

## Notes:

1. Applied voltage must be current limited to specified range.
2. Forcing voltage must be limited to specified range.
3. Current is specified as conventional current flowing into the device.

## Recommended Operating Conditions

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to Absolute Maximum Ratings.

| Symbol | Parameter |  | Min. | Nom. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $V_{\text {DD }}$ | Power Supply Voltage |  | 4.75 | 5.00 | 5.25 | V |
| $f_{s}$ | Conversion Rate | FMS3110 |  |  | 100 | Msps |
|  |  | FMS3115 |  |  | 150 |  |
| tpwh | CLK Pulse-width, HIGH | FMS3110 | 3.1 |  |  | ns |
|  |  | FMS3115 | 2.5 |  |  |  |
| $t_{\text {PWL }}$ | CLK Pulse-width, LOW | FMS3110 | 3.1 |  |  | ns |
|  |  | FMS3115 | 2.5 |  |  |  |
| tw | CLK Pulse-width | FMS3110 | 10 |  |  | ns |
|  |  | FMS3115 | 6.6 |  |  |  |
| $\mathrm{t}_{\text {s }}$ | Input Data Setup Time |  | 1.7 |  |  | ns |
| $\mathrm{t}_{\mathrm{h}}$ | Input Data Hold Time |  | 0 |  |  | ns |
| $\mathrm{V}_{\text {REF }}$ | Reference Voltage, External |  | 1.000 | 1.235 | 1.500 | V |
| $\mathrm{C}_{\mathrm{C}}$ | Compensation Capacitor |  |  | 0.1 |  | $\mu \mathrm{F}$ |
| $\mathrm{R}_{\text {LOAD }}$ | Output Load |  |  | 37.5 |  | $\Omega$ |
| $\mathrm{V}_{\mathrm{H}}$ | Input Voltage, Logic HIGH |  | 2.0 |  | $\mathrm{V}_{\mathrm{DD}}$ | V |
| $\mathrm{V}_{\text {IL }}$ | Input Voltage, Logic LOW |  | GND |  | 0.8 | V |
| $\mathrm{T}_{\mathrm{A}}$ | Ambient Temperature, Still Air |  | 0 |  | 70 | ${ }^{\circ} \mathrm{C}$ |

Electrical Characteristics

| Symbol | Parameter | Conditions ${ }^{(6)}$ | Min. | Typ. ${ }^{(4)}$ | Max. | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IDD | Power Supply Current ${ }^{(5)}$ | $\mathrm{V}_{\mathrm{DD}}=$ Max. |  |  | 125 | mA |
| $\mathrm{P}_{\mathrm{D}}$ | Total Power Dissipation ${ }^{(5)}$ | $\mathrm{V}_{\mathrm{DD}}=$ Max. |  |  | 655 | mW |
| Ro | Output Resistance |  |  | 100 |  | $\mathrm{k} \Omega$ |
| Co | Output Capacitance | $\mathrm{I}_{\text {OUT }}=0 \mathrm{~mA}$ |  |  | 30 | pF |
| $\mathrm{I}_{\mathrm{H}}$ | Input Current HIGH | $\mathrm{V}_{\mathrm{DD}}=$ Max., $\mathrm{V}_{\text {IN }}=2.4 \mathrm{~V}$ |  |  | -5 | $\mu \mathrm{A}$ |
| IIL | Input Current LOW | $\mathrm{V}_{\mathrm{DD}}=$ Max., $\mathrm{V}_{\text {IN }}=0.4$ |  |  | +5 | $\mu \mathrm{A}$ |
| $\mathrm{I}_{\mathrm{REF}}$ | $V_{\text {REF }}$ Input Bias Current |  |  | 0 | $\pm 100$ | $\mu \mathrm{A}$ |
| $\mathrm{V}_{\text {REF }}$ | Reference Voltage Output |  |  | 1.235 |  | V |
| Voc | Output Compliance | Referred to $\mathrm{V}_{\mathrm{DD}}$ | -0.4 | 0 | +1.5 | V |
| $\mathrm{C}_{\mathrm{DI}}$ | Digital Input Capacitance |  |  | 4 | 10 | pF |

Notes:
4. Values shown are typical for $\mathrm{V}_{\mathrm{DD}}=+5 \mathrm{~V}$ and $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
5. Minimum and Maximum values with $\mathrm{V}_{\mathrm{DD}}=\mathrm{Max}$ and $\mathrm{T}_{\mathrm{A}}=\mathrm{Min}$.
6. $V_{R E F}=1.235 \mathrm{~V}, R_{\text {LOAD }}=37.5 \Omega, \mathrm{R}_{\mathrm{REF}}=540 \Omega$.

## Switching Characteristics

| Symbol | Parameter | Conditions $^{(8)}$ | Min. | Typ. ${ }^{(7)}$ | Max. | Units |
| :---: | :--- | :--- | :--- | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{D}}$ | Clock to Output Delay | $\mathrm{V}_{\mathrm{DD}}=$ Min. |  | 10 | 15 | ns |
| $\mathrm{t}_{\text {SKEW }}$ | Output Skew |  |  | 1 | 2 | ns |
| $\mathrm{t}_{\mathrm{R}}$ | Output Rise Time | $10 \%$ to $90 \%$ of Full Scale |  |  | 3 | ns |
| $\mathrm{t}_{\mathrm{F}}$ | Output Fall Time | $90 \%$ to $10 \%$ of Full Scale |  |  | 3 | ns |

Notes:
7. Values shown are typical for $\mathrm{V}_{\mathrm{DD}}=+5 \mathrm{~V}$ and $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
8. $V_{R E F}=1.235 \mathrm{~V}, R_{\text {LOAD }}=37.5 \Omega, R_{\text {REF }}=590 \Omega$.

## System Performance Characteristics See Figure 3

| Symbol | Parameter | Conditions $^{(10)}$ | Min. | Typ. ${ }^{(\mathbf{9})}$ | Max. | Units |
| :---: | :--- | :--- | :--- | :---: | :---: | :---: |
| ELI | Integral Linearity Error | $\mathrm{V}_{\mathrm{DD}}, \mathrm{V}_{\mathrm{REF}}=$ Nominal |  | $\pm 0.10$ | $\pm 0.25$ | $\% / \mathrm{FS}$ |
| ELD | Differential Linearity Error | $\mathrm{V}_{\mathrm{DD}}, \mathrm{V}_{\mathrm{REF}}=$ Nominal |  | $\pm 0.10$ | $\pm 0.25$ | $\% / \mathrm{FS}$ |
| EDM | DAC to DAC Matching | $\mathrm{V}_{\mathrm{DD}}, \mathrm{V}_{\mathrm{REF}}=$ Nominal |  | 3 | 10 | $\%$ |
| PSRR | Power Supply Rejection Ratio |  |  |  | 0.05 | $\% / \%$ |

Notes:
9. Values shown are typical for $\mathrm{V}_{\mathrm{DD}}=+5 \mathrm{~V}$ and $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.
10. $\mathrm{V}_{\mathrm{REF}}=1.235 \mathrm{~V}$, $\mathrm{R}_{\mathrm{LOAD}}=37.5 \Omega$, $\mathrm{R}_{\mathrm{REF}}=590 \Omega$.

## Timing Information



Figure 3. Timing Diagram

## Equivalent Circuits



Figure 4. Equivalent Digital Input Circuit


Figure 5. Equivalent Analog Output Circuit


Figure 6. Equivalent Analog Input Circuit

## Functional Description

Within the FMS3110/3115 are three identical 10-bit D/A converters, each with a current source output. External loads are required to convert the current to voltage outputs. Data inputs RGB7-0 are overridden by the /BLANK input. /SYNC $=\mathrm{H}$ activates sync current from los for sync-on-green video signals.

## Digital Inputs

All digital inputs are TTL-compatible. Data is registered on the rising edge of the CLK signal. Following one stage of pipeline delay, the analog output changes too after the rising edge of CLK.

## ISYNC and /BLANK

ISYNC and /BLANK inputs control the output level (Figure 7 and Table 1) of the D/A converters during CRT retrace intervals. /BLANK forces the D/A outputs to the blanking level, while /SYNC = L turns off a current source connected to the green D/A converter. ISYNC $=\mathrm{H}$ adds a 40 IRE sync pulse to the green output; /SYNC $=\mathrm{L}$ sets the green output to 0.0 V during the sync tip. /SYNC and /BLANK are registered on the rising edge of CLK.
/BLANK gates the D/A inputs and sets the pedestal voltage. If /BLANK = HIGH, D/A inputs are added to a pedestal, which offsets the current output. If /BLANK = LOW, data inputs and the pedestal are disabled.


Figure 7. Nominal Output Levels

## DIA Outputs

Each D/A output is a current source. To obtain a voltage output, a resistor must be to ground. Output voltage depends upon this external resistor, the reference voltage, and the value of the gain-setting resistor connected between R $_{\text {REF }}$ and GND.

Normally, a $75 \Omega$ source termination resistor is connected between the D/A current output pin and GND near the D/A converter. A $75 \Omega$ line can be connected with another $75 \Omega$ termination resistor at the far end of the cable. This "double termination" presents the D/A converter a net resistive load of $37.5 \Omega$.

The FMS3110/3115 may also be operated with a single $75 \Omega$ terminating resistor. To lower the output voltage swing to the desired range, the nominal value of the resistor on $R_{\text {REF }}$ should be doubled.

## Voltage Reference

All three D/A converters are supplied with a common voltage reference. Internal bandgap voltage reference voltage is +1.235 V with a $3 \mathrm{~K} \Omega$ source resistance. An external voltage reference may be connected to the $\mathrm{V}_{\text {REF }}$ pin, overriding the internal voltage reference.

A $0.1 \mu \mathrm{~F}$ capacitor must be connected between the COMP pin and $V_{D D}$ to stabilize internal bias circuitry and ensure low-noise operation.

## Power and Ground

Required power is a single +5.0 V supply. To minimize power-supply induced noise, analog +5 V should be connected to $V_{D D}$ pins with 0.1 and $0.01 \mu \mathrm{~F}$ decoupling capacitors placed adjacent to each $V_{D D}$ pin or pin pair.
The high slew-rate of digital data makes capacitive coupling to the outputs of any D/A converter a potential problem. Since the digital signals contain high-frequency components of the CLK signal, as well as the video output signal, the resulting data feedthrough often looks like harmonic distortion or reduced signal-to-noise performance. All ground pins should be connected to a common solid ground plane for best performance.

Table 1. Output Voltage vs. Input Code, ISYNC and IBLANK ( $V_{R E F}=1.235 \mathrm{~V}, \mathrm{R}_{\mathrm{REF}}=590 \Omega, \mathrm{R}_{\mathrm{L}}=37.5 \Omega$ )

| RGB $_{9-0}$ (MSB...LSB) | Blue and Red DIAs |  |  | Green D/A |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | ISYNC | IBLANK | V $_{\text {out }}$ | ISYNC | /BLANK | V $_{\text {out }}$ |
| 1111111111 | X | 1 | 0.7140 | 1 | 1 | 1.0000 |
| 1111111111 | X | 1 | 0.7140 | 0 | 1 | 0.7140 |
| 1111111110 | X | 1 | 0.7134 | 1 | 1 | 0.9994 |
| 1111111101 | X | 1 | 0.7127 | 1 | 1 | 0.9987 |
| $:$ | $:$ | $:$ | $:$ | $:$ | $:$ | $:$ |
| 1000000000 | X | 1 | 0.3843 | 1 | 1 | 0.6703 |
| 0111111111 | X | 1 | 0.3837 | 1 | 1 | 0.6697 |
| $:$ | $:$ | $:$ | $:$ | $:$ | $:$ | $:$ |
| 0000000010 | X | 1 | 0.0553 | 1 | 1 | 0.3413 |
| 0000000001 | X | 1 | 0.0546 | 1 | 1 | 0.3406 |
| 0000000000 | X | 1 | 0.0540 | 1 | 1 | 0.3400 |
| XXXXXXXXX | X | 0 | 0.0000 | 1 | 0 | 0.2860 |
| XXXXXXXXXX | X | 0 | 0.0000 | 0 | 0 | 0.0000 |

## Application Information

Figure 8 illustrates a typical FMS3110/3115 interface circuit. In this example, an optional 1.2 V bandgap reference is connected to the $\mathrm{V}_{\text {REF }}$ output, overriding the internal voltage reference source.

## Grounding

It is important that the FMS3110/3115 power supply be well-regulated and free of high-frequency noise. Careful power supply decoupling ensures the highest quality video signals at the output of the circuit. The FMS3110/3115 has separate analog and digital circuits. To keep digital system noise from the D/A converter, it is recommended that power supply voltages ( $\mathrm{V}_{\mathrm{DD}}$ ) come from the system analog power source and all ground connections (GND) be made to the analog ground plane. Power supply pins should be individually decoupled at the pin.

## Printed Circuit Board Layout

Designing with high-performance mixed-signal circuits demands printed circuits with ground planes. Overall system performance is strongly influenced by the board layout. Capacitive coupling from digital to analog circuits may result in poor D/A conversion. Consider the following suggestions when designing the layout:

1. Keep the critical analog traces ( $\mathrm{V}_{\mathrm{REF}}, \mathrm{I}_{\mathrm{REF}}, \mathrm{COMP}$, los, lor, log) as short as possible and as far as possible from all digital signals. The FMS3110/3115 should be located near the board edge, close to the analog output connectors.
2. The power plane for the FMS3110/3115 should be separate from that which supplies the digital circuitry. A single power plane should be used for all of the $V_{D D}$ pins. If the power supply for the FMS3110/3115 is the same as that of the system's digital circuitry, power to the FMS3110/3115 should be decoupled with $0.1 \mu \mathrm{~F}$ and $0.01 \mu \mathrm{~F}$ capacitors and isolated with a ferrite bead.
3. The ground plane should be solid, not crosshatched. Connections to the ground plane should have very short leads.
4. If the digital power supply has a dedicated power plane layer, it should not be placed under the FMS3110/3115, the voltage reference, or the analog outputs. Capacitive coupling of digital power supply noise from this layer to the FMS3110/3115 and its related analog circuitry can have an adverse effect on performance.
5. CLK should be handled carefully. Jitter and noise on this clock degrade performance. Terminate the clock line carefully to eliminate overshoot and ringing.

## Related Products

- FMS38XX Triple 8-bit 150Msp D/A Converters
- FMS9884A 3x8-bit 140Ms/s A/D Converter


Figure 8. Typical Interface Circuit

## Physical Dimensions

| Symbol | Inches |  | Millimeters |  | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min. | Max. | Min. | Max. |  |
| A | . 055 | . 063 | 1.40 | 1.60 |  |
| A1 | . 001 | . 005 | . 05 | . 15 |  |
| A2 | . 053 | . 057 | 1.35 | 1.45 |  |
| B | . 006 | . 010 | . 17 | . 27 | 7 |
| D/E | . 346 | . 362 | 8.8 | 9.2 | 8 |
| D1/E1 | . 268 | . 284 | 6.8 | 7.2 | 2 |
| e | . 019 BSC |  | . 50 BSC |  |  |
| L | . 017 | . 029 | . 45 | . 75 | 6 |
| N | 48 |  | 48 |  | 4 |
| ND | 12 |  | 12 |  | 5 |
| $\alpha$ | 0 | 7 | 0 | 7 |  |
| cCC | . 004 |  | 0.08 |  |  |

Notes:

1. All dimensions and tolerances conform to ANSI Y14.5M-1982.
2. Dimensions "D1" and "E1" do not include mold protrusion. Allowable protrusion is 0.25 mm per side. D1 and E1 are maximum plastic body size dimensions including mold mismatch.
3. Pin 1 identifier is optional.
4. Dimension ND: Number of terminals.
5. Dimension ND: Number of terminals per package edge.
6. "L" is the length of terminal for soldering to a substrate.
7. Dimension " B " does not include dambar protrusion. Allowable dambar protrusion shall not cause the lead width to exceed the maximum B dimension by more than 0.08 mm . Dambar can not be located on the lower radius or the foot. Minimum space between protrusion and an adjacent lead is 0.07 mm for 0.4 mm and 0.5 mm pitch packages. To be determined at seating place $\tilde{N} \mathrm{CN}$
8. 

Figure 9. 48-Contact Leadless Quad Flat Package (LMQFP)
Package drawings are provided as a service to customers considering Fairchild components. Drawings may change in any manner without notice. Please note the revision and/or date on the drawing and contact a Fairchild Semiconductor representative to verify or obtain the most recent revision. Package specifications do not expand the terms of Fairchild's worldwide terms and conditions, specifically the warranty therein, which covers Fairchild products.

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| :---: | :---: | :---: | :---: |
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| Current Transfer Logic ${ }^{\text {TM }}$ | GTO ${ }^{\text {™ }}$ | Programmable Active Droop ${ }^{\text {TM }}$ | Pranche |
| Ecospark ${ }^{\text {® }}$ | $i-L O^{\text {TM }}$ | QFET ${ }^{\text {® }}$ | TinyBoost ${ }^{\text {m }}$ |
| EZSWITCH ${ }^{\text {TM }}$ * | Intellimax ${ }^{\text {m M }}$ | QS ${ }^{\text {TM }}$ | TinyBuck ${ }^{\text {TM }}$ |
| E1 ${ }^{\text {Im }}$ | ISOPLANAR ${ }^{\text {TM }}$ | QT Optoelectronics ${ }^{\text {TM }}$ | TinyLogic ${ }^{\text {® }}$ |
|  | MegaBuck ${ }^{\text {TM }}$ | Quiet Series ${ }^{\text {TM }}$ | TINYOPTOTM |
|  | MICROCOUPLER ${ }^{\text {TM }}$ | RapidConfigure ${ }^{\text {TM }}$ | TinyPower ${ }^{\text {TM }}$ |
| Fairchild ${ }^{\text {® }}$ | MicroFET ${ }^{\text {m }}$ | SMART START ${ }^{\text {TM }}$ | TinyPWM ${ }^{\text {m }}$ |
| Fairchild Semiconductor ${ }^{\text {® }}$ | MicroPak ${ }^{\text {m }}$ | SPM ${ }^{\text {® }}$ | TinyWire ${ }^{\text {TM }}$ |
| FACT Quiet Series ${ }^{\text {TM }}$ | MillerDrive ${ }^{\text {TM }}$ | STEALTH ${ }^{\text {TM }}$ | $\mu$ SerDes $^{\text {™ }}$ |
| FACT ${ }^{(1)}$ | Motion-SPM | SuperFET ${ }^{\text {m }}$ | UHC ${ }^{\text {® }}$ |
| FAST ${ }^{\text {® }}$ | OPTOLOGIC | SuperSOT ${ }^{\text {TM }}$-3 | Ultra FRFET ${ }^{\text {m }}$ |
| FastvCore ${ }^{\text {TM }}$ | OPTOPLANAR ${ }^{\text {® }}$ | SuperSOTTM 6 | UniFET ${ }^{\text {TM }}$ |
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## As used herein

1. Life support devices or systems are devices or systems which, (a) are intended for surgical implant into the body or (b) support or sustain life, and (c) whose failure to perform when properly used in accordance with instructions for use provided in the labeling, can be reasonably expected to result in a significant injury of the user.
2. A critical component in any component of a life support, device, or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

PRODUCT STATUS DEFINITIONS
Definition of Terms

| Datasheet Identification | Product Status | Definition |
| :--- | :--- | :--- |
| Advance Information | Formative or In Design | This datasheet contains the design specifications for product <br> development. Specifications may change in any manner without notice. |
| Preliminary | First Production | This datasheet contains preliminary data; supplementary data will be <br> published at a later date. Fairchild Semiconductor reserves the right to <br> make changes at any time without notice to improve design. |
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