FCAS50SN60 Smart Power Module for SRN

September 2006



# FCAS50SN60 Smart Power Module for SRM

## Features

- · Very low thermal resistance due to using DBC
- 600V-50A single-phase asymmetric bridge IGBT converter for SRM drive including control ICs for gate driving and protection
- Divided negative dc-link terminals for inverter current sensing applications
- Single-grounded power supply due to built-in HVIC
- Switching frequency of 2.2~8kHz
- Isolation rating of 2500Vrms/min.

# Applications

 AC 200V ~ 242V single-phase SRM drives for home application vacuum cleaner.

# **General Description**

FCAS50SN60 is an advanced smart power module for SRM drive that Fairchild has newly developed and designed to provide very compact and high performance SRM motor drives mainly targeting low-power inverter-driven SRM application especially for a vacuum air cleaner. It combines optimized circuit protection and drive matched to low-loss IGBTs. System reliability is further enhanced by the integrated under-voltage lock-out and short-circuit protection. The high speed built-in HVIC provides opto-coupler-less IGBT gate driving capability that further reduce the overall size of the inverter system design. In addition the incorporated HVIC facilitates the use of single-supply drive topology enabling the FCAS50SN60 to be driven by only one drive supply voltage without negative bias. Each phase current of inverter can be monitored separately due to the divided negative dc terminals.

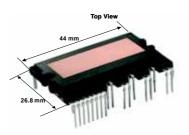




Figure 1.

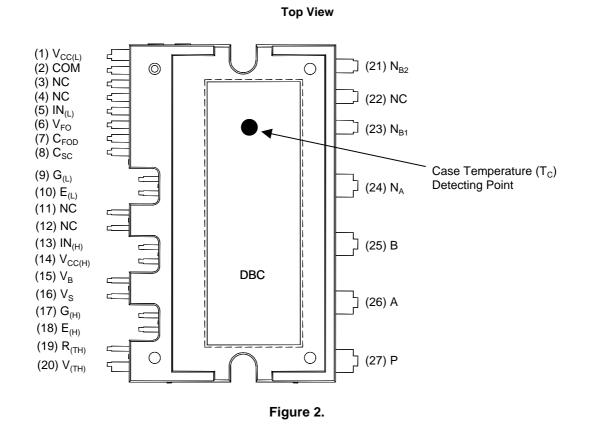
## **Integrated Power Functions**

• 600V-50A IGBT asymmetric converter for single-phase SRM drives (Please refer to Figure 3)

## Integrated Drive, Protection and System Control Functions

- For high-side IGBTs: Gate drive circuit, High voltage isolated high-speed level shifting Control circuit under-voltage (UV) protection Note) Available bootstrap circuit example is given in Figures 10.
- For low-side IGBTs: Gate drive circuit, Short circuit protection (SC) Control supply circuit under-voltage (UV) protection
- · Fault signaling: Corresponding to a UV fault (Low-side supply)
- Input interface: 5V CMOS/LSTTL compatible, Schmitt trigger input

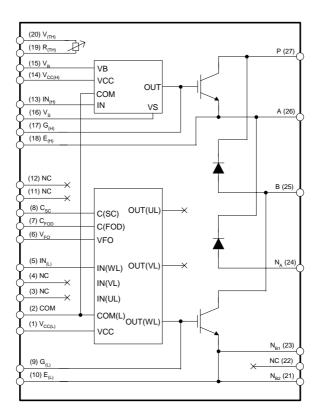
## **Pin Configuration**



Pin Number	Pin Name	Pin Description			
1	V <sub>CC(L)</sub>	Low-side Common Bias Voltage for IC and IGBTs Driving			
2	COM	Common Supply Ground			
3	NC	Dummy Pin			
4	NC	Dummy Pin			
5	IN <sub>(L)</sub>	Signal Input for Low-side IGBT			
6	V <sub>FO</sub>	Fault Output			
7	C <sub>FOD</sub>	pacitor for Fault Output Duration Time Selection			
8	C <sub>SC</sub>	pacitor (Low-pass Filter) for Short-Current Detection			
9	G <sub>(L)</sub>	Gate terminal of low-side IGBT			
10	E <sub>(L)</sub>	Emitter terminal of low-side IGBT			
11	NC	Dummy Pin			
12	NC	Dummy Pin			
13	IN <sub>(H)</sub>	Signal Input for High-side IGBT			
14	V <sub>CC(H)</sub>	High-side Bias Voltage			
15	VB	High-side Bias Voltage for Gate Driving			
16	Vs	High-side Bias Voltage Ground for Gate Driving			
17	G <sub>(H)</sub>	Gate terminal of the High-side IGBT			
18	E <sub>(H)</sub>	Emitter terminal of the High-side IGBT			
19	R <sub>(TH)</sub>	Thermistor Series Resistor			
20	V <sub>(TH)</sub>	Thermistor Bias Voltage			
21	N <sub>B2</sub>	Negative DC–Link Input for B Leg (Should be shorted with $N_{B1}$ externally)			
22	NC	Dummy Pin			
23	N <sub>B1</sub>	Negative DC–Link Input for B Leg (Should be shorted with N <sub>B2</sub> externally)			
24	N <sub>A</sub>	Negative DC-Link Input for A Leg			
25	В	Output for B Leg			
26	А	Output for A Leg			
27	Р	Positive DC–Link Input			

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# Internal Equivalent Circuit and Input/Output Pins



#### Note:

1. The low-side is composed of one IGBT and freewheeling diode and one control IC which has gate driving and protection functions.

2. The power side is composed of four dc-link input terminals and two output terminals.

3. The high-side is composed of one IGBT and freewheeling diode and one drive IC for high-side IGBT.

Figure 3.

# Absolute Maximum Ratings (T<sub>J</sub> = 25°C, Unless Otherwise Specified)

#### **Inverter Part**

Symbol	Parameter	Conditions	Rating	Units
V <sub>PN(Surge)</sub>	Supply Voltage (Surge)	Applied between P- N <sub>A</sub> , N <sub>B1</sub> , N <sub>B2</sub>	550	V
V <sub>CES</sub>	Collector-emitter Voltage		600	V
± I <sub>C</sub>	Each IGBT Collector Current	$T_{\rm C} = 25^{\circ}{\rm C}$	50	А
± I <sub>CP</sub>	Each IGBT Collector Current (Peak)	T <sub>C</sub> = 25°C, Under 1ms Pulse Width	100	А
P <sub>C</sub>	Collector Dissipation	T <sub>C</sub> = 25°C per One IGBT	110	W
ТJ	Operating Junction Temperature	(Note 1)	-20 ~ 125	°C

Note:

1. The maximum junction temperature rating of the power chips integrated within the module is 150 °C(@T<sub>C</sub> ≤ 100°C). However, to insure safe operation, the average junction temperature should be limited to T<sub>J(ave)</sub> ≤ 125°C (@T<sub>C</sub> ≤ 100°C)

## **Control Part**

Symbol	Parameter	Conditions	Rating	Units
V <sub>CC</sub>	Control Supply Voltage	Applied between V <sub>CC(H)</sub> , V <sub>CC(L)</sub> - COM	20	V
$V_{BS}$	High-side Control Bias Voltage	Applied between V <sub>B</sub> - V <sub>S</sub>	20	V
V <sub>IN</sub>	Input Signal Voltage	Applied between IN <sub>(H)</sub> , IN <sub>(L)</sub> - COM	-0.3~5.5	V
V <sub>FO</sub>	Fault Output Supply Voltage	Applied between V <sub>FO</sub> - COM	-0.3~V <sub>CC</sub> +0.3	V
I <sub>FO</sub>	Fault Output Current	Sink Current at V <sub>FO</sub> Pin	5	mA
V <sub>SC</sub>	Current Sensing Input Voltage	Applied between C <sub>SC</sub> - COM	-0.3~V <sub>CC</sub> +0.3	V

## **Total System**

Symbol	Parameter	Conditions	Rating	Units
V <sub>PN(PROT)</sub>	Self Protection Supply Voltage Limit (Short Circuit Protection Capability)	$V_{CC} = V_{BS} = 13.5 \sim 16.5 V$ T <sub>J</sub> = 125°C, Non-repetitive, less than 6µs	400	V
T <sub>C</sub>	Module Case Operation Temperature		-20 ~ 95	°C
T <sub>STG</sub>	Storage Temperature		-40 ~ 125	°C
V <sub>ISO</sub>	Isolation Voltage	60Hz, Sinusoidal, AC 1 minute, Connection Pins to DBC	2500	V <sub>rms</sub>

## **Thermal Resistance**

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Units
R <sub>th(j-c)Q</sub>	Junction to Case Thermal	Each IGBT under Operating Condition	-	-	0.90	°C/W
R <sub>th(j-c)F</sub>	Resistance	Each FWDi under Operating Condition	-	-	2.2	°C/W

Note:

2. For the measurement point of case temperature (T $_{C}$ ), please refer to Figure 2.

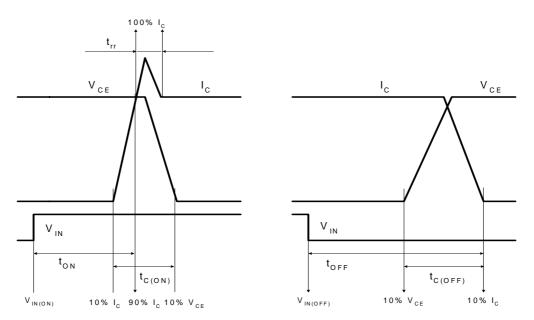
# Electrical Characteristics (T<sub>J</sub> = 25°C, Unless Otherwise Specified)

### **Inverter Part**

Sy	ymbol	Parameter	Condi	itions	Min.	Тур.	Max.	Units
V	CE(SAT)	Collector-Emitter Saturation Voltage	$\begin{array}{l} V_{CC} = V_{BS} = 15V \\ V_{IN} = 5V \end{array} \qquad \qquad I_C = 50A, \ T_J = 25^\circ C \end{array}$		-	1.6	2.3	V
	V <sub>FM</sub>	FWDi Forward Voltage	$V_{IN} = 0V$	$I_{\rm C} = 50$ A, $T_{\rm J} = 25^{\circ}$ C	-	2.1	3.0	V
HS	t <sub>ON</sub>	Switching Times	V <sub>PN</sub> = 300V, V <sub>CC</sub> = V <sub>BS</sub> = 15V		-	0.8	-	μS
	t <sub>C(ON)</sub>		$I_{C} = 50A$ $V_{IN} = 0V \leftrightarrow 5V$ , Inducti	ve Load	-	0.6	-	μS
	t <sub>OFF</sub>		$V_{IN} = 0.0 \iff 3.0$ , inductive Load R <sub>E(H)</sub> = 10 $\Omega$		-	1.5	-	μS
	t <sub>C(OFF)</sub>		(Note 3)		-	0.8	-	μS
	t <sub>rr</sub>				-	0.08	-	μS
LS	t <sub>ON</sub>		$V_{PN} = 300V, V_{CC} = V_{BS}$	<sub>S</sub> = 15V	-	1.1	-	μS
	t <sub>C(ON)</sub>		$I_{C} = 50A$ $V_{IN} = 0V \leftrightarrow 5V$ , Inducti	veload	-	0.9	-	μS
	t <sub>OFF</sub>		(Note 3)		-	1.5	-	μS
	t <sub>C(OFF)</sub>	]			-	0.8	-	μS
	t <sub>rr</sub>	]			-	0.05	-	μS
	I <sub>CES</sub>	Collector - Emitter Leakage Current	V <sub>CE</sub> = V <sub>CES</sub>		-	-	250	μA

#### Note:

3.  $t_{ON}$  and  $t_{OFF}$  include the propagation delay time of the internal drive IC.  $t_{C(ON)}$  and  $t_{C(OFF)}$  are the switching time of IGBT itself under the given gate driving condition internally. For the detailed information, please see Figure 4.





# **Electrical Characteristics** ( $T_J = 25^{\circ}C$ , Unless Otherwise Specified)

## **Control Part**

Symbol	Parameter	Cor	nditions	Min.	Тур.	Max.	Units
IQCCL	Quiescent V <sub>CC</sub> Supply Current	$V_{CC} = 15V$ $IN_{(L)} = 0V$	V <sub>CC(L)</sub> - COM	-	-	40	mA
I <sub>QCCH</sub>		$V_{CC} = 15V$ $IN_{(H)} = 0V$	V <sub>CC(H)</sub> - COM	-	-	80	μΑ
I <sub>QBS</sub>	Quiescent V <sub>BS</sub> Supply Current	V <sub>BS</sub> = 15V IN <sub>(H)</sub> = 0V	V <sub>B</sub> - V <sub>S</sub>	-	-	100	μΑ
V <sub>FOH</sub>	Fault Output Voltage	$V_{SC} = 0V, V_{FO}$ Circuit: 4.7k $\Omega$ to 5V Pull-up		4.5	-	-	V
V <sub>FOL</sub>		$V_{SC}$ = 1V, $V_{FO}$ Circuit: 4.7k $\Omega$ to 5V Pull-up		-	-	0.8	V
V <sub>SC(ref)</sub>	Short Circuit Trip Level	V <sub>CC</sub> = 15V (Note 4)		0.45	0.5	0.55	V
UV <sub>CCD</sub>	Supply Circuit Under-	Detection Level	Applied between V <sub>CC(L)</sub> - COM	10.5	-	12.5	V
UV <sub>CCR</sub>		Reset Level		11.0	-	13	V
UV <sub>BSD</sub>		Detection Level	Applied between	10.0	-	12.5	V
UV <sub>BSR</sub>		Reset Level	V <sub>B</sub> - V <sub>S</sub>	10.5	-	13.0	V
t <sub>FOD</sub>	Fault-out Pulse Width	C <sub>FOD</sub> = 33nF (Note 5)		1.4	1.8	2.0	ms
V <sub>IH</sub>	ON Threshold Voltage	Logic'1' input voltage	Applied between	3.0	-	-	V
V <sub>IL</sub>	OFF Threshold Voltage	Logic'0' input voltage	IN <sub>(H)</sub> , IN <sub>(L)</sub> - COM	-	-	0.8	V
I <sub>INH(ON)</sub>	Input Bias Current	IN <sub>(H)</sub> = 5V	Applied between	0.9	-	2.2	mA
I <sub>INL(ON)</sub>		$IN_{(L)} = 5V$	IN <sub>(H)</sub> , IN <sub>(L)</sub> - COM	0.9	-	2.4	mA
R <sub>TH</sub>	Resistance of Thermistor	@ T <sub>C</sub> = 25°C (Note Fig. 10)		-	50	-	kΩ
		@ T <sub>C</sub> = 80°C (Note Fi	g. 10)	-	5.76	-	kΩ

Note:

4. Short-circuit current protection is functioning only at the low-sides.

5. The fault-out pulse width  $t_{FOD}$  depends on the capacitance value of  $C_{FOD}$  according to the following approximate equation :  $C_{FOD} = 18.3 \times 10^{-6} \times t_{FOD}[F]$ 

# **Recommended Operating Conditions**

Symbol	Parameter	Conditions	Value			Unito
Symbol	Symbol Farameter	Conditions	Min.	Тур.	Max.	Units
V <sub>PN</sub>	Supply Voltage	Applied between P - N <sub>A</sub> , N <sub>B1</sub> , N <sub>B2</sub>	-	300	450	V
V <sub>CC</sub>	Control Supply Voltage	Applied between V <sub>CC(H)</sub> , V <sub>CC(L)</sub> - COM	13.5	15	16.5	V
V <sub>BS</sub>	High-side Bias Voltage	Applied between V <sub>B</sub> - V <sub>S</sub>	13.5	15	18.5	V
f <sub>PWM</sub>	PWM Input Signal	$T_{C} \le 100^{\circ}C, T_{J} \le 125^{\circ}C$	-	3	-	kHz
V <sub>IN(ON)</sub>	Input ON Voltage	Applied between IN <sub>(H)</sub> , IN <sub>(L)</sub> - COM		4 ~ 5.5		V
V <sub>IN(OFF)</sub>	Input OFF Voltage	Applied between IN <sub>(H)</sub> , IN <sub>(L)</sub> - COM		0 ~ 0.65		V

# Mechanical Characteristics and Ratings

Parameter	6	Conditions			Limits		
Farameter					Max.	Units	
Mounting Torque	Mounting Screw - M3		5.17	6.29	7.30	Kg•cm	
			0.51	0.62	0.72	N•m	
Surface Flatness		Note Figure 5.	0	-	120	um	
Weight			-	15.0	-	g	

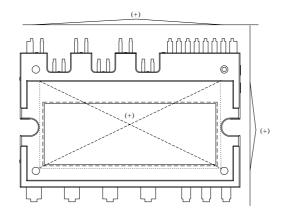
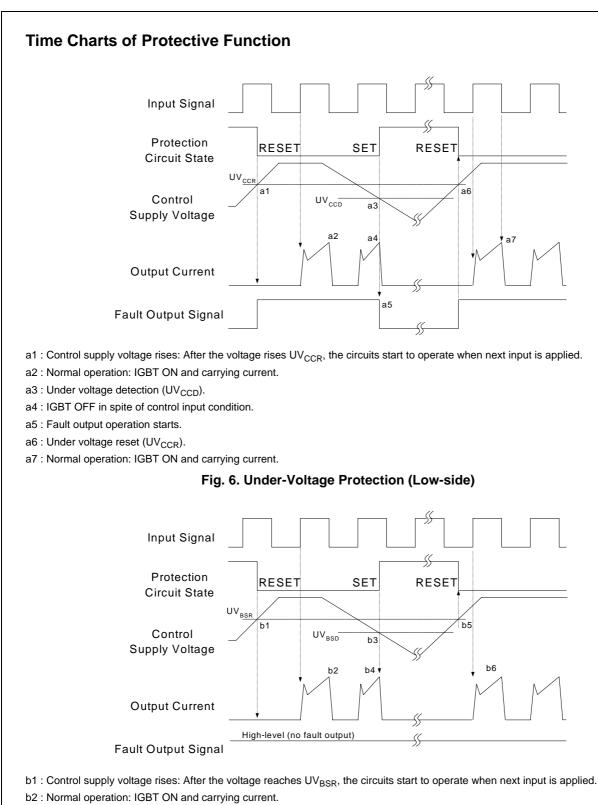
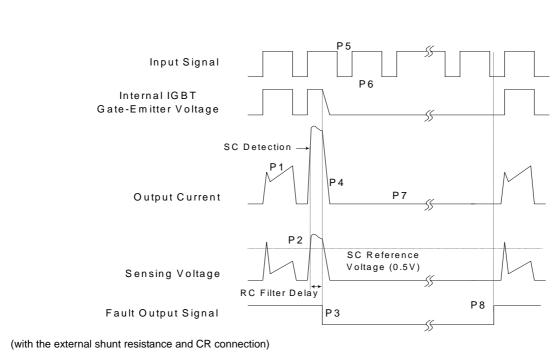


Figure 5. Flatness Measurement Position



- b3 : Under voltage detection (UV<sub>BSD</sub>).
- b4 : IGBT OFF in spite of control input condition, but there is no fault output signal.
- b5 : Under voltage reset (UV<sub>BSR</sub>)
- b6 : Normal operation: IGBT ON and carrying current

#### Fig. 7. Under-Voltage Protection (High-side)



c1 : Normal operation: IGBT ON and carrying current.

c2 : Short circuit current detection (SC trigger).

c3 : Hard IGBT gate interrupt.

c4 : IGBT turns OFF.

c5 : Fault output timer operation starts: The pulse width of the fault output signal is set by the external capacitor C<sub>FO</sub>.

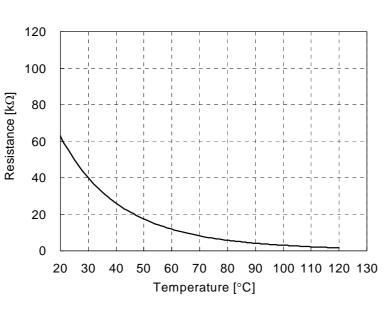
c6 : Input "L" : IGBT OFF state.

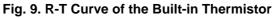
c7 : Input "H": IGBT ON state, but during the active period of fault output the IGBT doesn't turn ON.

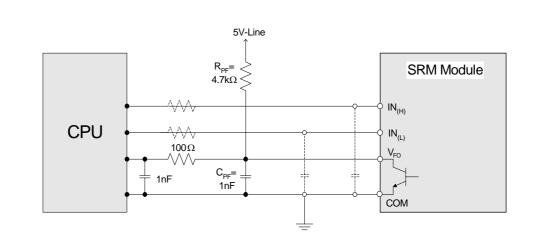
c8 : IGBT OFF state

## Fig. 8. Short-Circuit Current Protection (Low-side Operation only)

**R-T Graph** 



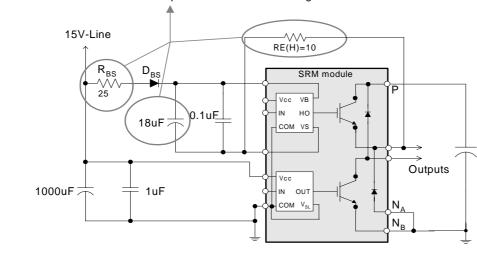




#### Note:

- RC coupling at each input (parts shown dotted) might change depending on the PWM control scheme used in the application and the wiring impedance of the application's printed circuit board. The input signal section integrates 3.3kΩ(typ.) pull-down resistor. Therefore, when using an external filtering resistor, please pay attention to the signal voltage drop at input terminal.
- 2. The logic input is compatible with standard CMOS or LSTTL outputs.

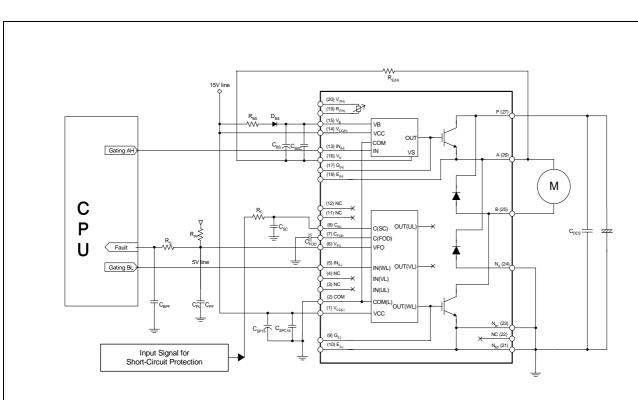
#### Figure 10. Recommended CPU I/O Interface Circuit



This Value depends on PWM Control Algorithm

#### Note:

It would be recommended that the bootstrap diode, D<sub>BS</sub>, has soft and fast recovery characteristics. R<sub>BS</sub> should be 2.5 times greater than R<sub>E(H)</sub> **Figure 11. Recommended Bootstrap Operation Circuit and Parameters** 

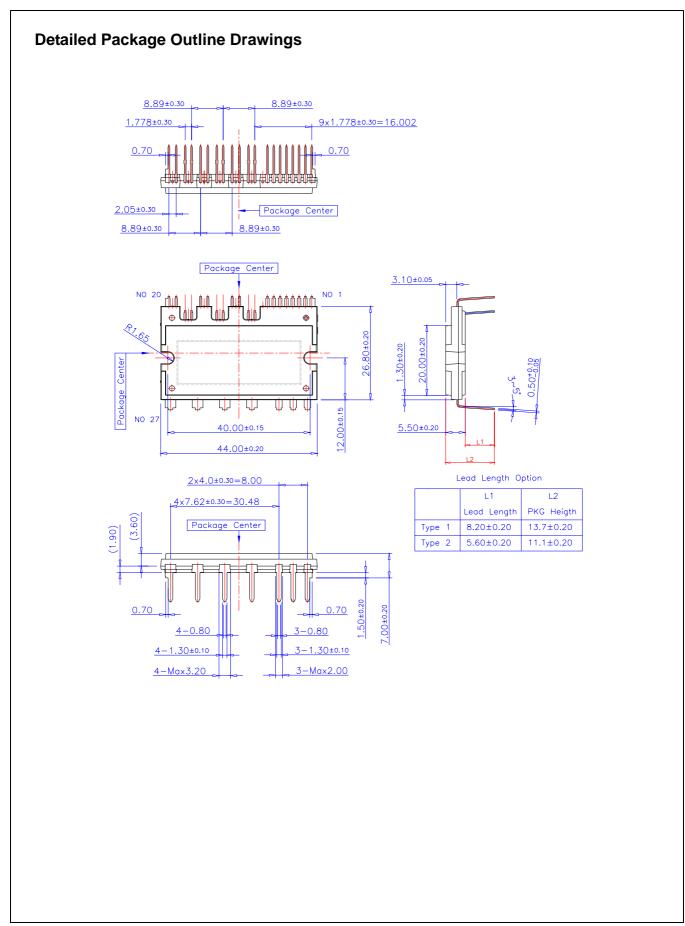


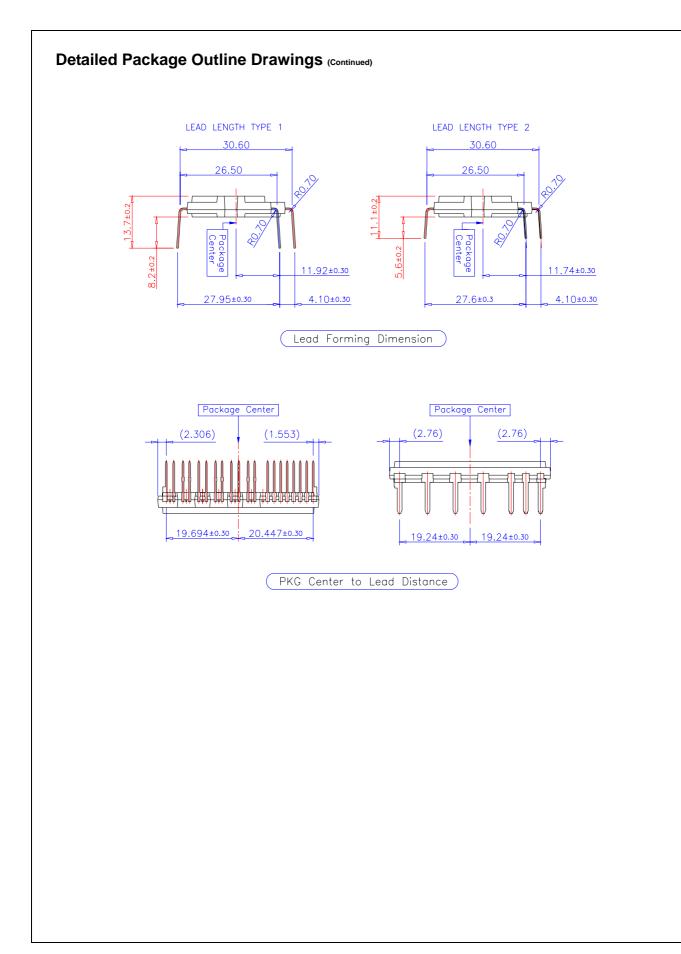
#### Note:

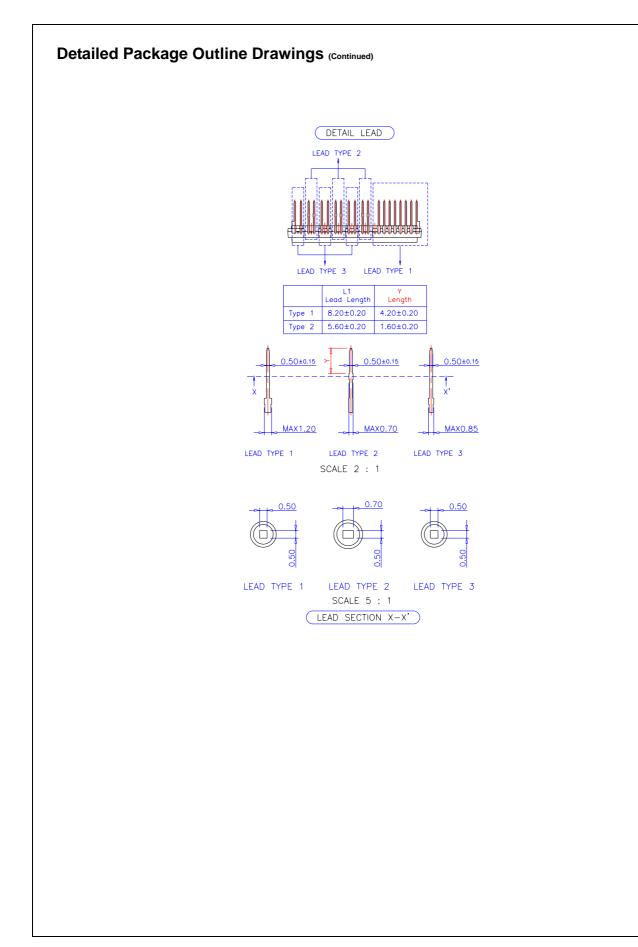
- 1. To avoid malfunction, the wiring of each input should be as short as possible. (less than 2-3cm)
- 2. By virtue of integrating an application specific type HVIC inside the Module, direct coupling to CPU terminals without any opto-coupler or transformer isolation is possible.
- V<sub>FO</sub> output is open collector type. This signal line should be pulled up to the positive side of the 5V power supply with approximately 4.7kΩ resistance. Please refer to Figure 10.
- 4.  $C_{\text{SP15}}$  of around 7 times larger than bootstrap capacitor  $C_{\text{BS}}$  is recommended.
- 5. V<sub>FO</sub> output pulse width should be determined by connecting an external capacitor(C<sub>FOD</sub>) between C<sub>FOD</sub>(pin7) and COM(pin2). (Example : if C<sub>FOD</sub> = 33 nF, then t<sub>FO</sub> = 1.8ms (typ.)) Please refer to the note 6 for calculation method.
- 6. Input signal is High-Active type. There is a 3.3kΩ resistor inside the IC to pull down each input signal line to GND. When employing RC coupling circuits, set up such RC couple that input signal agree with turn-off/turn-on threshold voltage
- 7. To prevent errors of the protection function, the wiring around R<sub>SC</sub>, R<sub>F</sub> and C<sub>SC</sub> should be as short as possible.
- 8. In the short-circuit protection circuit, please select the  $R_FC_{SC}$  time constant in the range 3~4  $\mu s.$
- 9. Each capacitor should be mounted as close to the pins as possible.
- 10. To prevent surge destruction, the wiring between the smoothing capacitor and the P&N pins should be as short as possible. The use of a high frequency non-inductive capacitor of around 0.1~0.22 µF between P and N pins is recommended.
- 11. Relays are used at almost every systems of electrical equipments of home appliances. In these cases, there should be sufficient distance between the CPU and the relays.
- 12. C<sub>SPC15</sub> should be over 1uF and mounted as close to the pins of the module as possible.

13. N<sub>B1</sub>(pin23) and N<sub>B2</sub>(pin21) should be shorted externally.

### Fig. 12. Application Circuit







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